

編號: 277 系所: 電腦與通信工程研究所丙組

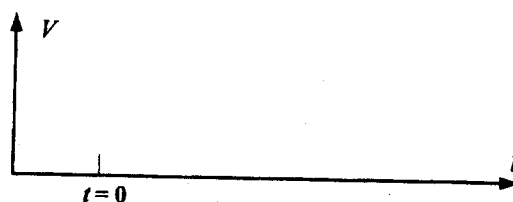
科目: 電磁學及電磁波

本試題是否可以使用計算機:  可使用,  不可使用 (請命題老師勾選)\* Useful constants:  $\epsilon_0 = 10^{-9}/36\pi$ ;  $\mu_0 = 4\pi \times 10^{-7}$ ;  $\eta_0 = 120\pi$ 

1. (a) For a **perfectly-matched two-port microwave 3-dB attenuator (3-dB 衰減器)**, determine its S-parameter matrix. (5%)
- (b) If the **input and output VSWR** of a two-port 3-dB microwave attenuator are all equal to **1.5** (and assuming the input and output impedance are all **real**), determine its S-parameter matrix. (5%)

$$\text{S-parameter matrix: } \begin{bmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{bmatrix}$$

2. At  $t = 0$  for an **unit step signal** incident to a **short-circuited** microstrip transmission line with a **length  $L$**  on a substrate of **thickness  $d$**  and **dielectric constant  $\epsilon_r$** , draw the **reflected** and **total signal voltage** as a **function of time**. (Note: assuming a pure TEM wave propagating in the microstrip transmission line) (10%)



3. Determine the length  $l$  (in term of wavelength  $\lambda$ ) of a **short-circuited** transmission line ( $Z_0 = 50 \Omega$ ) to have an **equivalent inductance  $L = 5 \text{ nH}$**  at  $f = 1 \text{ GHz}$ . (10%)

Note: assuming the transmission line is lossless with a **propagation constant  $\beta$** .

\* General input impedance formula of a lossless transmission line with a length  $l$  and load  $Z_L$ : 
$$Z_i = Z_0 \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)}$$

4. For an air-filled rectangular waveguide ( $a \times b$ ), it can be proven that the normalized frequency at which the  $\text{TE}_{10}$  mode exhibits its minimum attenuation is

$$f/f_c = \left[ g + \sqrt{g^2 - (2b/a)} \right]^{1/2} \quad g = \frac{3}{2} + \frac{3b}{a}$$

If a waveguide is with ( $a = 6 \text{ mm}$ ,  $b = 3 \text{ mm}$ ), determine the operating frequency of the  $\text{TE}_{10}$  mode which exhibits its minimum attenuation. (10%)

Waveguide  $\text{TE}_{10}$  mode attenuation constant :

$$\alpha_{c\text{TE}_{10}} = \frac{1}{\eta_0 b} \sqrt{\frac{\pi f \mu_0}{\sigma_c [1 - (f_c/f)^2]}} \left[ 1 + 2 \frac{b}{a} (f_c/f)^2 \right]$$

(背面仍有題目, 請繼續作答)

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5. The Helmholtz's equation in a **source-free lossy medium** ( $\epsilon, \mu, \sigma$ ) is as follows.

$$\nabla^2 \bar{E} + k_c^2 \bar{E} = 0$$

$$k_c = \sqrt{-j\omega\mu(\sigma + j\omega\epsilon)} = \omega\sqrt{\mu\epsilon\left(1 + \frac{\sigma}{j\omega\epsilon}\right)} = \omega\sqrt{\mu\epsilon_c}$$

$\epsilon_c =$  equivalent complex - permittivity

- (a) Determine the approximated **intrinsic (wave) impedance**  $\eta_c$  of a TEM wave in a **good conductor** ( $\sigma/\epsilon\omega \gg 1$ ). (10%)

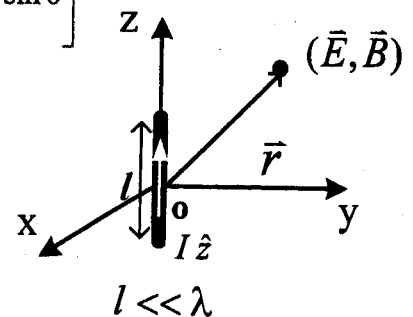
Note: Use the condition  $\sigma/\epsilon\omega \gg 1$  and  $\sqrt{j} = \frac{(1+j)}{\sqrt{2}}$

- (b) From (a) explain that E- and H-field of a TEM wave in a good conductor have a **phase difference of 45°**. (5%)

6. For a **z-directed Hertzian dipole antenna** with an **uniform current distribution**  $I$  in a free space, the **radiated E- and B-field** can be derived as follows.

$$\bar{E} = \frac{-j\eta_0 I l}{4\pi k} \left[ \hat{r} \left( \frac{2}{r^3} + \frac{j2k_0}{r^2} \right) e^{-jk_0 r} \cos\theta + \hat{\theta} \left( \frac{1}{r^3} + \frac{jk_0}{r^2} - \frac{k_0^2}{r} \right) e^{-jk_0 r} \sin\theta \right]$$

$$\bar{B} = \hat{\phi} \frac{\mu_0 I l (1 + jk_0 r)}{4\pi r^2} e^{-jk_0 r} \sin\theta$$



- (a) Derive the **near-zone and far-zone radiated E- and H-field**. (10%)
- (b) From (a) Express the relation between the **far-zone E- and H-field**. (5%)
- (c) Draw an approximated **E-plane** (at  $\phi = 0^\circ$ ) antenna pattern in **linear scale**. (5%)
- (d) Determine the far-zone **radiation power density**  $P_{av}$  and the **total radiation power**  $P_{rad}$ . (15%)
- Note:  $\left( \int_0^\pi (\sin\theta)^3 d\theta = 4/3 \right)$
- (e) Explain the meaning of the **antenna radiation resistance**  $R_{rad}$ . Determine the  $R_{rad}$  of a Hertzian dipole. (10%)